

TECH NOTE :: ClipX as PID controller

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Brief description

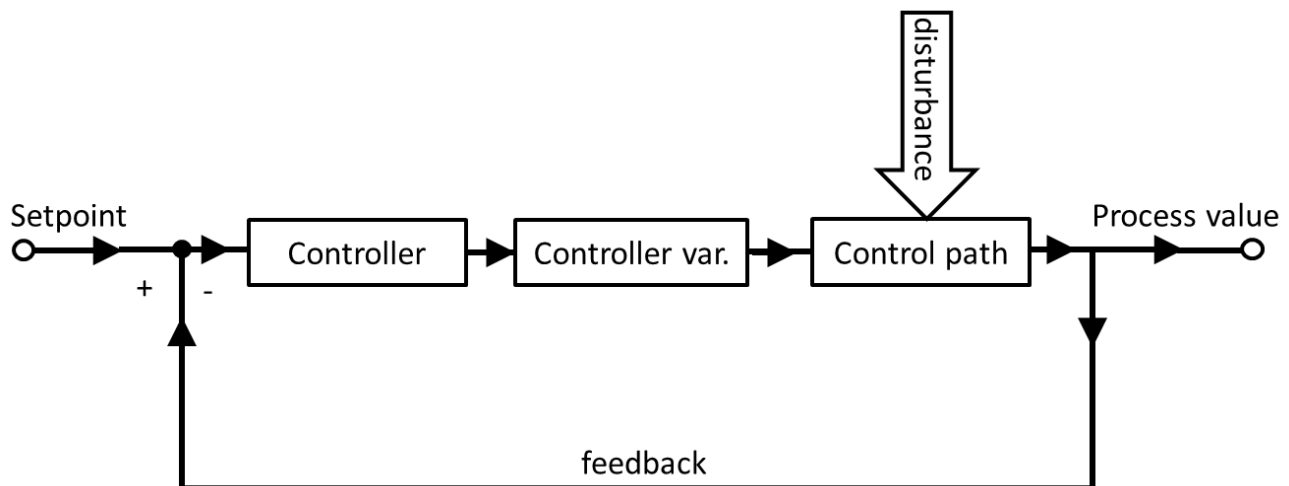
This is an instruction for using ClipX as PID controller using the internal calculated channels. In this example the charge process of a simple RC circuit first order is controlled with the PID controller. Furthermore, the effects of a sudden unloading on the control and the regulating value are shown.

1. Theory

Control engineering in general

Basics

When controlling a physical value, a feedback of the process value is used to react to disturbance variables to force the actual value to follow a setpoint value. This behavior is far superior compared to a simple controller that cannot react to any changes within the system. The block diagram of a control loop is as follows:



The setpoint provides a value that is compared to the process value, the actual value. The difference between setpoint value and process value results in a control variable that affects the control path. That way disturbance values should be countered to maintain the set point value in a certain range. Besides accuracy, in practice a control loop has also to be adjusted in terms of speed and stability.

PID controller

The PID controller (proportional-integral-derivative controller) consists of a P-, I-, and D-controller. The basic characteristics of the different types are quickly pointed out below:

P-controller:

The process value is proportionally adjusted to the control variable. This adjustment is time-independent and essentially depends on the proportionality gain K_P . The control deviation is minimalized when using a relatively high proportionality factor, though results in oscillation of the system. That is why a compromise has to be found when choosing the proportionality factor. For that reason the P-controller always has a certain control deviation.

The P-controller works fast but imprecisely. As a well-known example of this is the radiator with a thermostatic valve (P controller).

I-controller:

The I-controller is capable of adjusting the process value very precisely to the setpoint by adding the time integral of the control deviation to the control variable. If the control deviation is greater than zero a balance is never achieved.

The I-controller indeed is very precise, however it takes a lot of time for the integration and therefore is quite slow.

D-controller:

By differentiating, the D-controller generates a control variable that is proportional to the gradient of the control deviation. That way a high counterbalance for a quickly changing control deviation is possible. For the case that the control deviation is a constant value the D-controller does not provide any input because the gradient is equal zero.

RC circuit controlling

Components

The RC circuit used consists of a 56kΩ resistor that is connected in parallel to a 120μF electrolytic capacitor.

The time constant is the result of the product of the resistance and the capacitance:

$$\begin{aligned}\tau &= R \cdot C \\ &= 56\text{k}\Omega \cdot 120\mu\text{F} \\ &= 6,72\text{s}\end{aligned}$$

In practice a whole charging process of a capacitor (about 99%) takes approximately the time of five times the time constant:

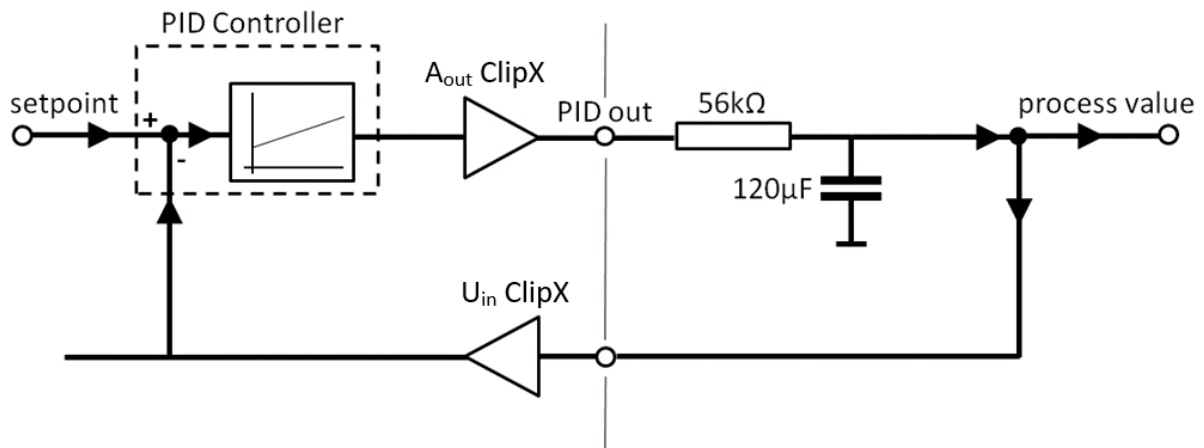
$$\begin{aligned}t_L &= 5 \cdot \tau \\ &= 5 \cdot 6,72\text{s} \\ &= 33,6\text{s}\end{aligned}$$

Important Note: Because the components have a certain tolerances they should be verified for sure. The usage of the correct time constant is very important for adjusting the PID-controller.

2. Carrying out

Connecting

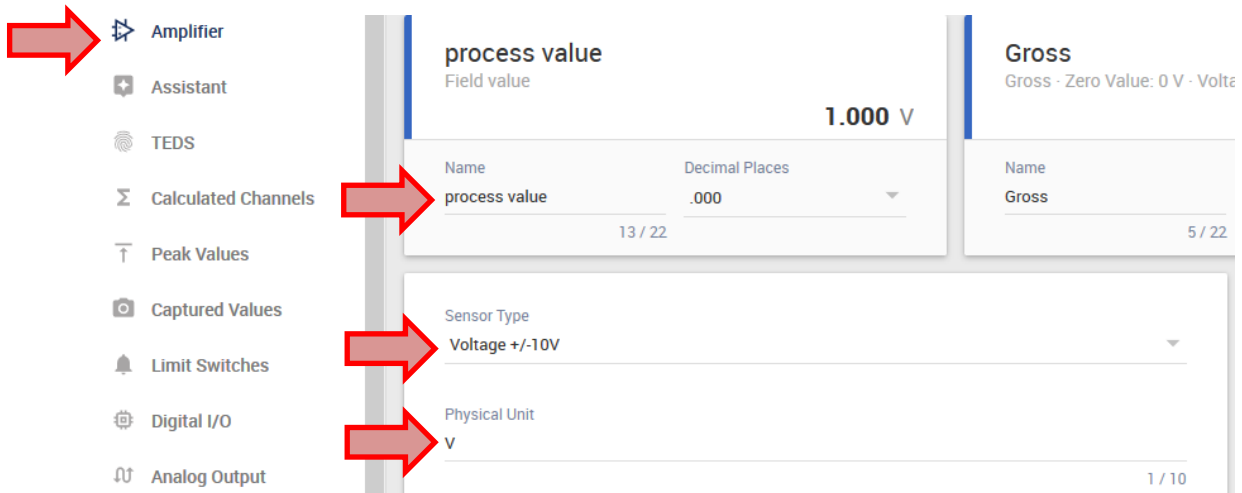
First, the RC circuit must be connected to the ClipX like shown below. Therefore, connect the positive input path of the RC circuit (see marker of the E capacitor) to the positive analog output and the output path to the positive voltage input of ClipX. The other path is connected to ground on both sides.



ClipX settings

Amplifier settings

- Switch to the menu item 'Amplifier'
- Select 'Voltage +/-10V' as sensor type
- Set the physical unit to 'V'
- Adapt the channel name to 'process value'

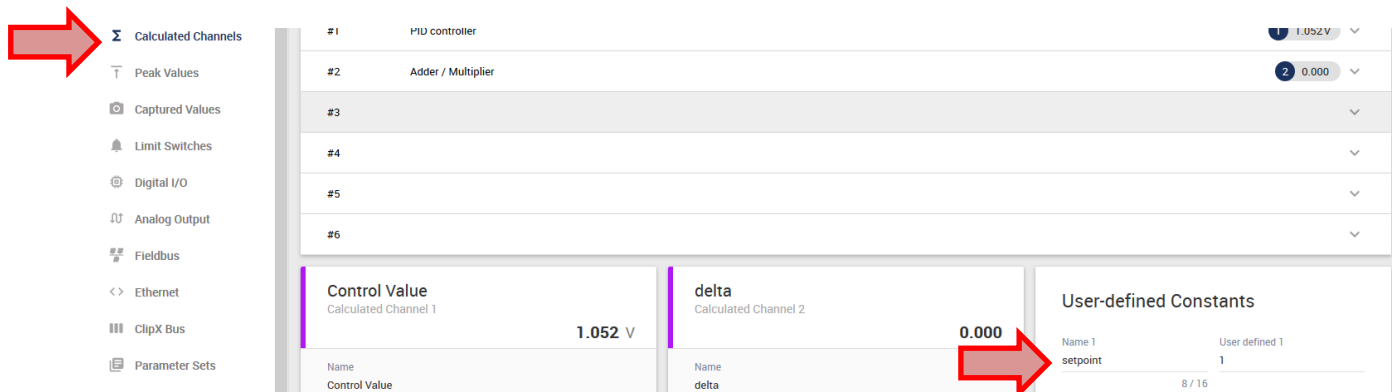


Calculated channels

Setpoint value:

The setpoint is defined as a constant signal.

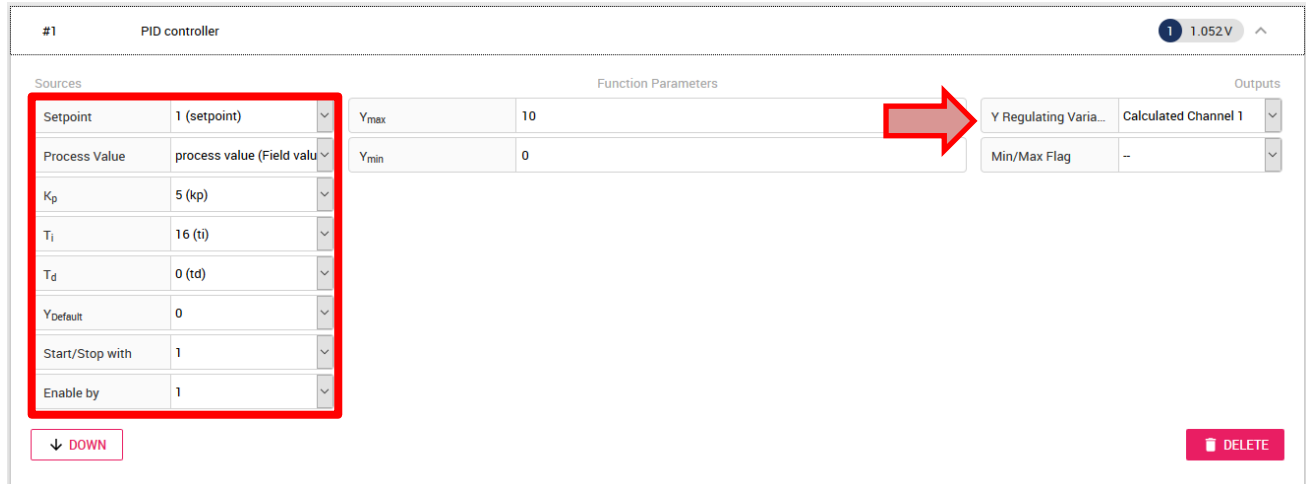
- Go to 'Calculated Channels'
- Create a user defined constant named 'setpoint' with a fitting value on the right side



PID controller:

Now the pid controller is set up.

- In the menu 'calculated channels' add a new channel of type 'PID controller'
- Select the created constant as setpoint
- Select the voltage input as process value
- Set up the control parameters (further explanation below)
- Assign the regulating value to a calculated channel



Sources		Function Parameters		Outputs	
Setpoint	1 (setpoint)	Y _{max}	10	Y Regulating Varia...	Calculated Channel 1
Process Value	process value (Field valu	Y _{min}	0	Min/Max Flag	--
K _p	5 (kp)				
T _i	16 (ti)				
T _d	0 (td)				
Y _{default}	0				
Start/Stop with	1				
Enable by	1				

↓ DOWN

DELETE

All parameters of the PID controller are shortly described here:

K_p: The proportionality-gain specifies the amplification from the P-controller

T_i [s]: The integral-time of the I-controller which is mainly depended on the time constant τ (if not used set $T_i = \infty$; $PMX: T_i = 1e38$).

T_d [s]: Derivative-time is the time the D-controller needs to differentiate. In this example set to 0, because no counterbalance is needed ($T_d = 0$).

Y_{max} und Y_{min}: The adjustment range, ie the maximum controller output in both directions.

Assigning the regulating/control value to the analog output of ClipX

- Switch to 'Analog Output'
- Select 'Voltage +/-10V' as type
- Select the control/regulating value as source

Amplifier

Assistant

TEDS

Calculated Channels

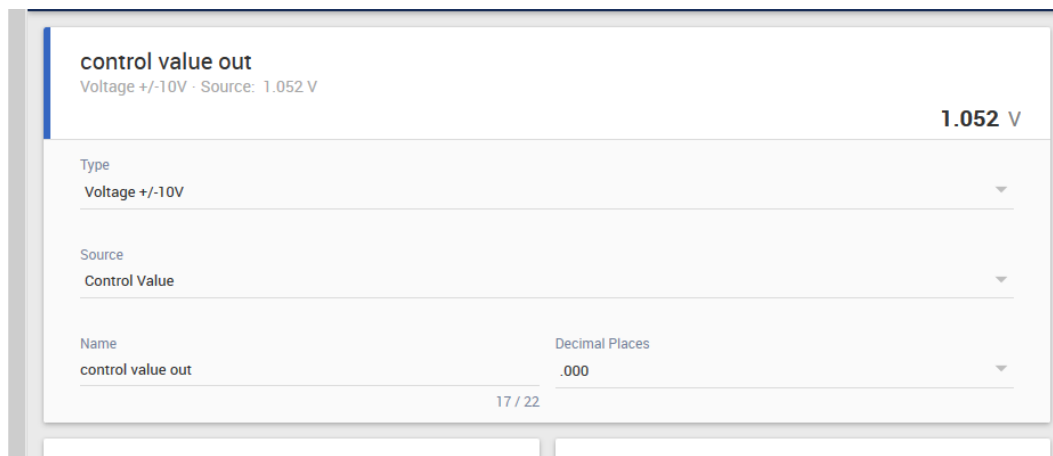
Peak Values

Captured Values

Limit Switches

Digital I/O

Analog Output



control value out
Voltage +/-10V · Source: 1.052 V

1.052 V

Type	Source	Name	Decimal Places
Voltage +/-10V	Control Value	control value out	.000

17 / 22

Control deviation:

The control deviation results from the difference between process value and setpoint value. With the help of a simple adder that difference can be displayed.

Note: This calculated channel is optional and not necessary for the PID controller. It is only created as an information source.

- Go to 'Calculated Channels'
- Add a 'Adder/Multiplier'
- Assign the process value to x1
- Set x2, x3 and x4 to 1
- Assign the setpoint to x5
- Set x6 to -1
- Assign the result to a calculated channel

#2
Adder / Multiplier
2 0.000 ^

$$y = x_1 x_2 x_3 x_4 + x_5 x_6 + x_7 x_8 + x_9 x_{10}$$

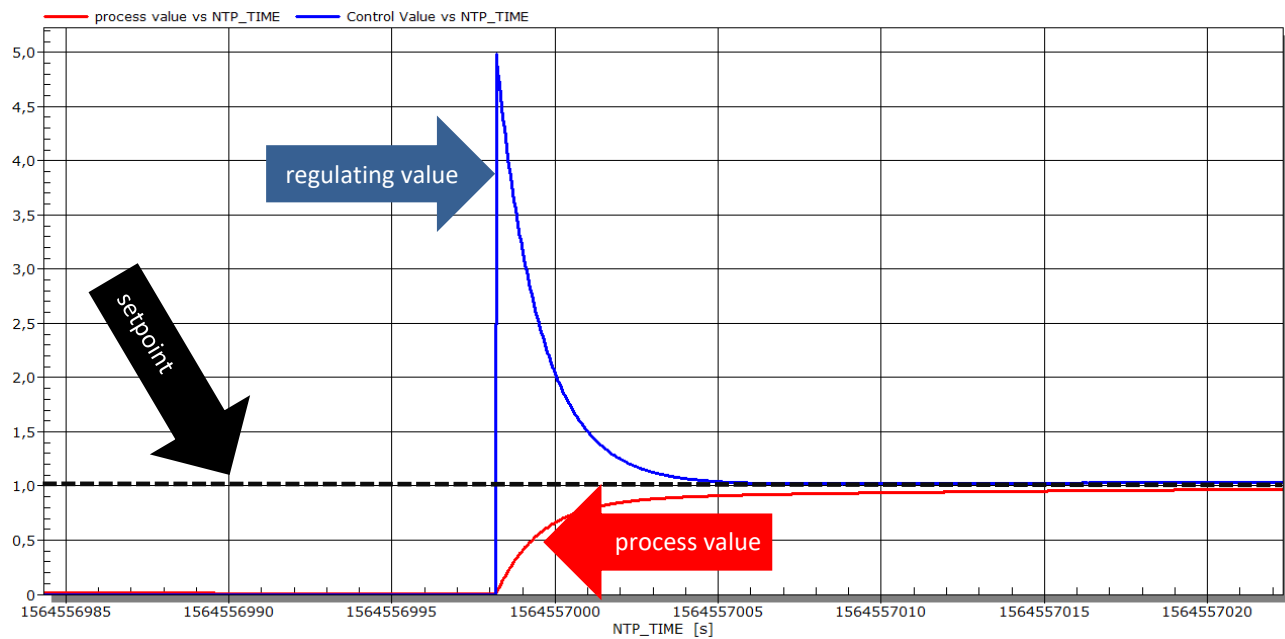
x ₁	process value (Field value)	x ₂	1	x ₃	1
x ₄	1	x ₅	1 (setpoint)	x ₆	-1
x ₇	0	x ₈	0	x ₉	0
x ₁₀	0				

y
Calculated Channel 2

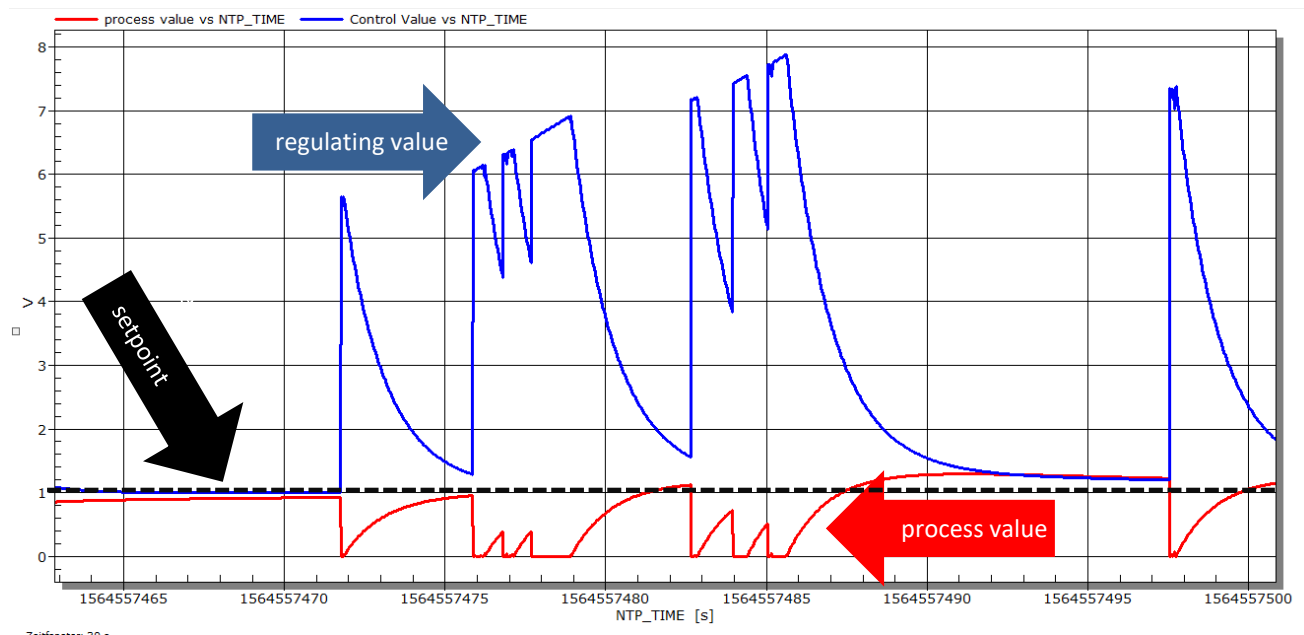
UP DOWN
DELETE

Graphical presentation

Charging curve of the capacitor



Interference by shortly discharging the capacitor



Determine control parameters

First Ziegler-Nichols method

The method of Ziegler and Nichols is a heuristic procedure to determine the control parameters proportionality-gain (K_p), integral-time (T_i) and derivative-time (T_d).

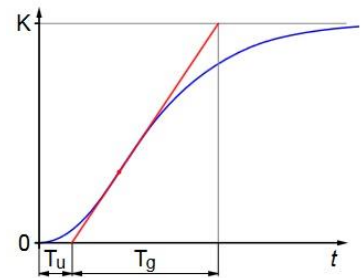
The approach is pretty easy: First the two parameters proportionality gain (K_p) and derivative-time (T_d) of the PID controller are set to zero and the integral-time (T_i) is set to ∞ (PMX: $T_i=1e38$). Then the proportionality gain (K_p) of the P-controller is increased until the stability limit of the system is reached. At that maximum amplitude K_u the system oscillates with constant amplitude. With the parameters K_u and the period of oscillation T_u the parameters of the PID controller are now adjusted.

1. Z-N tuning	K_p	T_i	T_d
P controller	$0.5 K_u$	-	-
PI controller	$0.45 K_u$	$0.85 T_u$	-
PD controller	$0.55 K_u$	-	$0.15 T_u$
PID classic Ziegler-Nichols	$0.6 K_u$	$0.5 T_u$	$0.125 T_u$
PID Pessen Integral rule	$0.7 K_u$	$0.4 T_u$	$0.15 T_u$
PID Some Overshoot	$0.33 K_u$	$0.5 T_u$	$0.33 T_u$
PID No Overshoot	$0.2 K_u$	$0.5 T_u$	$0.33 T_u$

Second Ziegler-Nichols method

In the second approach the controlling path is approximated as transfer element first order with dead time (PT1T_t-element). The stationary amplification K_s , the time constant T and the dead time T_t have to be known. Those can be found out by approximation with the help of the step function response, so that the following can be applied:

$$T = T_g \text{ und } T_t = T_u.$$



2. Z-N tuning	K_p	T_i	T_d
P controller	$\frac{1}{K_s} \cdot \frac{T}{T_t}$	-	-
PI controller	$0.9 \cdot \frac{1}{K_s} \cdot \frac{T}{T_t}$	$3.33 \cdot T_t$	-
PID controller	$\frac{1}{K_s} \cdot \frac{T}{T_t}$	$2 \cdot T_t$	$0.5 \cdot T_t$

Hint: Classic Z-N tuning is designed for countering disturbances fast and quickly and therefore results in some overshoot. Furthermore the control parameters have to be determined experimentally at the uncontrolled process, which is the reason to be very careful that it takes no damage.

Chien, Hrones and Reswick tuning rules

The method of Chien, Hrones and Reswick can be used for control paths of a higher order and is an advancement of the second Ziegler-Nichols method. The differentiation into periodic and aperiodic controlling as well as the separated settings for set-point-value and disturbance-value are beneficial.

The parameters stationary amplification K_s , time constant T and dead time T_t have to be known and can also be found out with the step function response, if the process is suitable for that.

The calculation is as shown in the table.

Chien, Hrones and Reswick tuning		Aperiodic			Periodic (20% overshoot)		
		K_p	T_i	T_d	K_p	T_i	T_d
P controller	Disturbance	$0.3 \cdot \frac{T_g}{T_u \cdot K_s}$	-	-	$0.7 \cdot \frac{T_g}{T_u \cdot K_s}$	-	-
	FÜHRUNG	$0.3 \cdot \frac{T_g}{T_u \cdot K_s}$	-	-	$0.7 \cdot \frac{T_g}{T_u \cdot K_s}$	-	-
PI controller	Disturbance	$0.6 \cdot \frac{T_g}{T_u \cdot K_s}$	$4 \cdot T_u$	-	$0.7 \cdot \frac{T_g}{T_u \cdot K_s}$	$2.3 \cdot T_u$	-
	FÜHRUNG	$0.35 \cdot \frac{T_g}{T_u \cdot K_s}$	$1.2 \cdot T_g$	-	$0.6 \cdot \frac{T_g}{T_u \cdot K_s}$	$1 \cdot T_g$	-
PID controller	Disturbance	$0.95 \cdot \frac{T_g}{T_u \cdot K_s}$	$2.4 \cdot T_u$	$0.42 \cdot T_u$	$1.2 \cdot \frac{T_g}{T_u \cdot K_s}$	$2 \cdot T_u$	$0.42 \cdot T_u$
	FÜHRUNG	$0.6 \cdot \frac{T_g}{T_u \cdot K_s}$	$1 \cdot T_g$	$0.5 \cdot T_u$	$0.95 \cdot \frac{T_g}{T_u \cdot K_s}$	$1.35 \cdot T_g$	$0.47 \cdot T_u$

Disclaimer

These examples are for illustrative purposes only. They cannot be used as the basis for any warranty or liability claims.